







SN65HVD30 – SN65HVD35

SLLS665I - SEPTEMBER 2005-REVISED APRIL 2010

3.3 V FULL-DUPLEX RS-485 DRIVERS AND RECEIVERS

Check for Samples: SN65HVD30 - SN65HVD35

FEATURES

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- 1/8 Unit-Load Option Available (Up to 256 Nodes on the Bus)
- Bus-Pin ESD Protection Exceeds 15 kV HBM
- Optional Driver Output Transition Times for Signaling Rates⁽¹⁾ of 1 Mbps, 5 Mbps and 26 Mbps
- Low-Current Standby Mode: < 1 μA
- Glitch-Free Power-Up and Power-Down Protection for Hot-Plugging Applications
- 5-V Tolerant Inputs
- . Bus Idle, Open, and Short Circuit Failsafe
- Driver Current Limiting and Thermal Shutdown
- Designed for RS-422 and RS-485 Networks
- 5-V Devices available, SN65HVD50-55
- (1) Line Signaling Rate is the number of voltage transitions made per second expressed in units of bps (bits per second).

APPLICATIONS

- Utility Meters
- DTE/DCE Interfaces
- Industrial, Process, and Building Automation
- Point-of-Sale (POS) Terminals and Networks

DESCRIPTION

The SN65HVD3X devices are 3-state differential line drivers and differential-input line receivers that operate with 3.3-V power supply.

Each driver and receiver has separate input and output pins for full-duplex bus communication designs. They are designed for RS-422 and RS-485 data transmission over cable lengths of up to 1500 meters.

The SN65HVD30, SN65HVD31, and SN65HVD32 are fully enabled with no external enabling pins.

The SN65HVD33, SN65HVD34, and SN65HVD35 have active-high driver enables and active-low receiver enables. A low, less than $1\mu A,$ standby current can be achieved by disabling both the driver and receiver.

All devices are characterized for ambient temperatures from -40°C to 85°C. Low power dissipation allows operation at temperatures up to 105°C or 125°C, depending on package option.

IMPROVED REPLACEMENT FOR:

Part Number	Replace with	
xxx3491 xxx3490	SN65HVD33: SN65HVD30:	Better ESD protection (15kV vs 2kV or not specified) Higher Signaling Rate (26Mbps vs 20Mbps) Fractional Unit Load (64 Nodes vs 32)
MAX3491E MAX3490E	SN65HVD33: SN65HVD30:	Higher Signaling Rate (26Mbps vs 12Mbps) Fractional Unit Load (64 Nodes vs 32)
MAX3076E MAX3077E	SN65HVD33: SN65HVD30:	Higher Signaling Rate (26Mbps vs 16Mbps) Lower Standby Current (1 μA vs 10 μA)
MAX3073E MAX3074E	SN65HVD34: SN65HVD31:	Higher Signaling Rate (5Mbps vs 500kbps) Lower Standby Current (1 μ A vs 10 μ A)
MAX3070E MAX3071E	SN65HVD35: SN65HVD32:	Higher Signaling Rate (1Mbps vs 250kbps) Lower Standby Current (1 μ A vs 10 μ A)



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



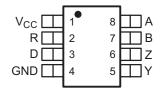


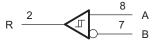
This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

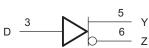
ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

SN65HVD30, SN65HVD31, SN65HVD32

D PACKAGE (TOP VIEW)

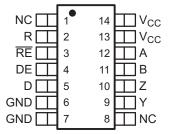




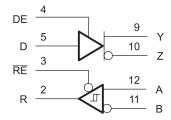


SN65HVD33, SN65HVD34, SN65HVD35

D PACKAGE (TOP VIEW)

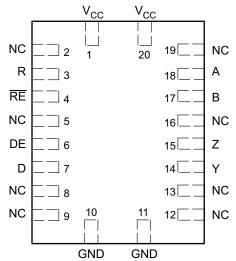


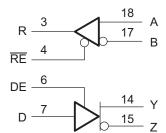
NC - No internal connection
Pins 6 and 7 are connected together internally
Pins 13 and 14 are connected together internally



SN65HVD33

RHL PACKAGE (TOP VIEW)





NC - No internal connection

Pins 10 and 11 are connected together internally Pins 1 and 20 are connected together internally



AVAILABLE OPTIONS

SIGNALING RATE	UNIT LOADS	ENABLES	BASE PART NUMBER	SOIC MARKING
26 Mbps	1/2	No	SN65HVD30	VP30
5 Mbps	1/8	No	SN65HVD31	VP31
1 Mbps	1/8	No	SN65HVD32	VP32
26 Mbps	1/2	Yes	SN65HVD33	65HVD33
5 Mbps	1/8	Yes	SN65HVD34	65HVD34
1 Mbps	1/8	Yes	SN65HVD35	65HVD35

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range unless otherwise noted (1) (2)

		UNIT
V _{CC}	Supply voltage range	−0.3 V to 6 V
$V_{(A)}, V_{(B)}, V_{(Y)}, V_{(Z)}$	Voltage range at any bus terminal (A, B, Y, Z)	−9 V to 14 V
V _(TRANS)	Voltage input, transient pulse through 100 Ω . See Figure 12 (A, B, Y, Z) ⁽³⁾	–50 to 50 V
V _I	Input voltage range (D, DE, RE)	-0.5 V to 7 V
Io	Output current (receiver output only, R)	11 mA

⁽¹⁾ Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

DISSIPATION RATINGS

PACKAGE	JEDEC THERMAL MODEL	T _A < 25°C RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 85°C RATING	T _A = 105°C RATING	T _A = 125°C RATING
SOIC (D) 9 pip	Low k	625 mW	5 mW/°C	325 mW		
SOIC (D) 8 pin	High k	1000 mW	8 mW/°C	520 mW	360 mW	
COIC (D) 14 pin	Low k	765 mW	6.1 mW/°C	400 mW	275 mW	
SOIC (D) 14 pin	High k	1350 mW	10.8 mW/°C	705 mW	485 mW	270 mW
QFN (RHL) 20 pin	High k	1710 mW	13.7 mW/°C	890 mW	6150 mW	340 mW

⁽²⁾ All voltage values, except differential I/O bus voltages, are with respect to network ground terminal.

⁽³⁾ This tests survivability only and the output state of the receiver is not specified.



RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range unless otherwise noted

				MIN	NOM	MAX	UNIT	
V_{CC}	Supply voltage			3		3.6	V	
V _I or V _{IC}	Voltage at any bus	s terminal (sep	arately or common mode)	-7 ⁽¹⁾		12	V	
		SN65HVD30	, SN65HVD33			26		
1/t _{UI}	Signaling rate	SN65HVD31	, SN65HVD34			5	5 Mbps	
		SN65HVD32	, SN65HVD35		1 54 60 2 V _{CC}			
R_{L}	Differential load re	sistance		54	60		Ω	
V _{IH}	High-level input vo	oltage	D, DE, RE	2		V_{CC}		
V _{IL}	Low-level input vo	ltage	D, DE, RE	0		0.8	V	
V _{ID}	Differential input v	oltage		-12		12		
	High lovel evitorit	araat	Driver	-60			m Λ	
ГОН	High-level output of	current	Receiver	-8			mA	
	Lavidaval avitavit a		Driver			60	A	
I _{OL}	Low-level output c	urrent	Receiver			8	mA	
TJ	Junction temperate	ure		-40		150	°C	

⁽¹⁾ The algebraic convention, in which the least positive (most negative) limit is designated as minimum is used in this data sheet.

ELECTROSTATIC DISCHARGE PROTECTION

PARAMETER	TEST CONDITIONS	MIN TYP ⁽¹⁾ MAX	UNIT
Human body model	Bus terminals and GND	±16	
Human body model (2)	All pins	±4	kV
Charged-device-model ⁽³⁾	All pins	±1	

- All typical values at 25°C with 3.3-V supply.
 Tested in accordance with JEDEC Standard 22, Test Method A114-A.
 Tested in accordance with JEDEC Standard 22, Test Method C101.



DRIVER ELECTRICAL CHARACTERISTICS

over recommended operating conditions unless otherwise noted

	PARAMETER	1	TEST CONDIT	IONS	MIN	TYP ⁽¹⁾	MAX	UNIT	
V _{I(K)}	Input clamp voltage		$I_I = -18 \text{ mA}$	$I_I = -18 \text{ mA}$				V	
			$I_O = 0$		2.5		V_{CC}		
D. (Steady-state differential output voltage		$R_L = 54 \Omega$, See Figure 1 (RS	-485)	1.5	2		V	
$ V_{OD(SS)} $			$R_L = 100 \Omega$, See Figure 1, (2)) (RS-422)	2	2.3			
			$V_{\text{test}} = -7 \text{ V to } 12 \text{ V, See Figure } V_{\text{test}} = -7 \text{ V to } 12 \text{ V}$	ure 2	1.5				
$\Delta V_{OD(SS)} $	Change in magnitud differential output vo states		$R_L = 54 \Omega$, See Figure 1 and	Figure 2	-0.2		0.2	V	
V _{OD(RING)}	Differential Output V and undershoot	oltage overshoot	$R_L = 54 \Omega$, $C_L = 50 pF$, See Figure 3	Figure 5 and			10% ⁽³⁾	V	
	Peak-to-peak	HVD30, HVD33				0.5			
$V_{OC(PP)}$	common-mode output voltage	HVD31, HVD34, HVD32, HVD35	See Figure 4			0.25		V	
V _{OC(SS)}	Steady-state commo voltage	on-mode output	Con Firms 4				2.3	V	
$\Delta V_{OC(SS)}$	Change in steady-st output voltage	ate common-mode	See Figure 4		-0.05		0.05	V	
		HVD30, HVD31,	$V_{CC} = 0 \text{ V}, V_Z \text{ or } V_Y = 12 \text{ V},$ Other input at 0 V				90		
$I_{Z(Z)}$ or	High-impedance	HVD32	$V_{CC} = 0 \text{ V}, V_{Z} \text{ or } V_{Y} = -7 \text{ V},$ Other input at 0 V		-10				
$I_{Y(Z)}$	state output current	HVD33, HVD34,	V _{CC} = 3 V or 0 V, DE = 0 V V _Z or V _Y = 12 V	Other input			90	μΑ	
		HVD35	$V_{CC} = 3 \text{ V or } 0 \text{ V, DE} = 0 \text{ V}$ $V_Z \text{ or } V_Y = -7 \text{ V}$	at 0 V	-10				
I _{Z(S)} or	Ohant Cinavit avitavit		V_Z or $V_Y = -7 \text{ V}$	Other input	-250		250	A	
I _{Y(S)}	Short Circuit output current ⁽⁴⁾		V_Z or $V_Y = 12 V$	at 0 V	-250		250	mA	
I _I	Input current	D, DE			0		100	μА	
C _(OD)	Differential output ca	apacitance	$V_{OD} = 0.4 \sin (4E6\pi t) + 0.5 V$	/, DE at 0 V		16		pF	

⁽¹⁾ All typical values are at 25°C and with a 3.3-V supply.

⁽²⁾ V_{CC} is 3.3 $Vdc \pm 5\%$

^{(3) 10%} of the peak-to-peak differential output voltage swing, per TIA/EIA-485

⁽⁴⁾ Under some conditions of short-circuit to negative voltages, output currents exceeding the ANSI TIA/EIA-485-A maximum current of 250 mA may occur. Continuous exposure may affect device reliability. This applies to the HVD30,HVD31,HVD33, and HVD34.



DRIVER SWITCHING CHARACTERISTICS

over recommended operating conditions unless otherwise noted

	PARAMETER		TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT	
		HVD30, HVD33		4	10	18		
t _{PLH}	Propagation delay time, low-to-high-level output	HVD31, HVD34		25	38	65	ns	
	odipat	HVD32, HVD35		120	175	305		
		HVD30, HVD33		4	9	18		
t _{PHL}	Propagation delay time, high-to-low-level output	HVD31, HVD34		25	38	65	ns	
	odipat	HVD32, HVD35		120	175	305		
		HVD30, HVD33		2.5	5	12		
t _r	Differential output signal rise time	HVD31, HVD34	$R_L = 54 \Omega, C_L = 50 pF,$ See Figure 5	20	37	60	ns	
		HVD32, HVD35	Occ rigulo o	120	185	300		
		HVD30, HVD33		2.5	5	12		
t _f	Differential output signal fall time	HVD31, HVD34		20	35	60	ns	
		HVD32, HVD35		120	180	300		
		HVD30, HVD33			0.6			
t _{sk(p)}	Pulse skew (t _{PHL} - t _{PLH})	HVD31, HVD34		2.0		ns		
		HVD32, HVD35			5.1	5.1		
		HVD33				45		
	Propagation delay time, high-impedance-to-high-level output	HVD34	D 440 0 DE ++ 0 \/			235	ns	
	riigh-impedance-to-nigh-level output	HVD35	$R_L = 110 \Omega$, \overline{RE} at 0 V, D = 3 V and S1 = Y, or			490		
		HVD33	D = 0 V and S1 = Z			25		
t _{PHZ}	Propagation delay time, high-level-to-high-impedance output	HVD34	See Figure 6			65	ns	
	riigh-level-to-high-impedance output	HVD35				165		
		HVD33				35		
t _{PZL1}	Propagation delay time, high-impedance-to-low-level output	HVD34	D 440 0 DE ++ 0 \/			190	ns	
	riigh impedance to low level output	HVD35	$R_L = 110 \Omega$, \overline{RE} at 0 V, D = 3 V and S1 = Z, or			490		
		HVD33	D = 0 V and S1 = Y			30		
t_{PLZ}	Propagation delay time, low-level-to-high-impedance output	HVD34	See Figure 7			120	ns	
	low-level-to-nigh-impedance output	HVD35		29		290		
t _{PZH1,} t _{PZL1}	Driver enable delay with bus voltage offse	et	V _O = 2 V (Typ)		500	900	ns	
t _{PZH2}	Propagation delay time, standby-to-high-l	$R_L = 110 \ \Omega, \overline{RE} \ at 3 \ V,$ $D = 3 \ V \ and \ S1 = Y, \ or$ $D = 0 \ V \ and \ S1 = Z$ See Figure 6			4000	ns		
t _{PZL2}	Propagation delay time, standby-to-low-le	vel output	$R_L = 110 \ \Omega$, \overline{RE} at 3 V, D = 3 V and S1 = Z, or D = 0 V and S1 = Y See Figure 7			4000	ns	

⁽¹⁾ All typical values are at 25°C and with a 3.3-V supply.



RECEIVER ELECTRICAL CHARACTERISTICS

over recommended operating conditions unless otherwise noted

	PARAMETE	:R	TEST CONDITION	ONS	MIN	TYP ⁽¹⁾	MAX	UNIT
V _{IT+}	Positive-going differential voltage	input threshold	$I_O = -8 \text{ mA}$				-0.02	V
V _{IT-}	Negative-going differentia voltage	I input threshold	I _O = 8 mA		-0.20			V
V_{hys}	Hysteresis voltage (V _{IT+} -	V _{IT-})				50		mV
V_{IK}	Enable-input clamp voltag	је	$I_{I} = -18 \text{ mA}$		-1.5			V
\/			$V_{ID} = 200 \text{ mV}, I_{O} = -8 \text{ mA}, Se$	ee Figure 8	2.4			V
Vo	Output voltage		$V_{ID} = -200 \text{ mV}, I_{O} = 8 \text{ mA}, Se$	ee Figure 8			0.4	V
$I_{O(Z)}$	High-impedance-state out	tput current	$V_O = 0$ or V_{CC} , \overline{RE} at V_{CC}	$V_O = 0$ or V_{CC} , \overline{RE} at V_{CC}			1	μА
		HVD31, HVD32,	V_A or $V_B = 12 \text{ V}$	Other input at 0V		0.05	0.1	mA
			V_A or $V_B = 12 \text{ V}$, $V_{CC} = 0 \text{ V}$			0.06	0.1	
		HVD34, HVD35	V_A or $V_B = -7 V$		-0.10	-0.04		
I _A or	Due input ourrent		V_A or $V_B = -7 V$, $V_{CC} = 0 V$		-0.10	-0.03		
IB	Bus input current		V_A or $V_B = 12 V$			0.20	0.35	
		LIV/D20 LIV/D22	V_A or $V_B = 12 \text{ V}$, $V_{CC} = 0 \text{ V}$	Other input at		0.24	0.4	
		HVD30, HVD33	V_A or $V_B = -7 \text{ V}$	0V	-0.35	-0.18		mA
			V_A or $V_B = -7 \text{ V}$, $V_{CC} = 0 \text{ V}$		-0.25	-0.13		
I _{IH}	Input current, RE		V _{IH} = 0.8 V or 2 V		-60			μА
C_{ID}	Differential input capacita	nce	$V_{ID} = 0.4 \sin (4E6\pi t) + 0.5 V,$	DE at 0 V		15		pF

⁽¹⁾ All typical values are at 25°C and with a 3.3-V supply.

SUPPLY CURRENT CHARACTERISTICS

over recommended operating conditions unless otherwise noted

	PARA	METER	TEST CONDITIONS	MIN TYP(1)	MAX	UNIT	
		HVD30	Dat 0 V as V and No Load		2.1	A	
		HVD31, HVD32	D at 0 V or V _{CC} and No Load	3.8	6.4	mA	
		HVD33	RE at 0 V, D at 0 V or V _{CC} , DE at 0 V,		1.8	m Λ	
Icc	HVD34, HVD35 No load (Receiver enabled and driver disabled)		2.2	mA			
	Supply current	HVD33, HVD34, HVD35	$\overline{\text{RE}}$ at V_{CC} , D at V_{CC} , DE at 0 V, No load (Receiver disabled and driver disabled)	0.022	1	μА	
		HVD33	RE at 0 V, D at 0 V or V _{CC} , DE at V _{CC} ,		2.1	mA	
		HVD34, HVD35	No load (Receiver enabled and driver enabled)		6.5		
		HVD33	RE at V _{CC} , D at 0 V or V _{CC} , DE at V _{CC}		1.8	mA	
		HVD34, HVD35	No load (Receiver disabled and driver enabled)		6.2		

⁽¹⁾ All typical values are at 25°C and with a 3.3-V supply.



RECEIVER SWITCHING CHARACTERISTICS

over recommended operating conditions unless otherwise noted

	PARA	METER	TEST (CONDITIONS	MIN TYP ⁽¹⁾	MAX	UNIT
	Propagation delay time,	HVD30, HVD33			26	45	
t _{PLH}	low-to-high-level output	HVD31, HVD32, HVD34, HVD35			47	70	
	Propagation delay time,	HVD30, HVD33			29	45	
t _{PHL}	high-to-low-level output	HVD31, HVD32, HVD34, HVD35	V _{ID} = -1.5 V	to 1.5 V,	49	70	
	Dulas show (It the I)	HVD30, HVD33	$C_L = 15 \text{ pF},$	See Figure 9		7	
t _{sk(p)}	Pulse skew (t _{PHL} - t _{PLH})	HVD31, HVD34, HVD32, HVD35				10	
t _r	Output signal rise time					5	
t _f	Output signal fall time					6	ns
t _{PHZ}	Output disable time from hig	gh level	DE at 3 V			20	
t _{PZH1}	Output enable time to high I	evel	DE at 3 V	$C_L = 15 \text{ pF},$ See Figure 10		20	
t _{PZH2}	Propagation delay time, star	ndby-to-high-level output	DE at 0 V	Occ rigare to		4000	
t _{PLZ}	Output disable time from lov	v level	DE -+ 2.1/	_		20	
t _{PZL1}	Output enable time to low le	evel	DE at 3 V	C _L = 15 pF, See Figure 11		20	
t _{PZL2}	Propagation delay time, star	ndby-to-low-level output	DE at 0 V	- Coorigulo 11		4000	

⁽¹⁾ All typical values are at 25°C and with a 3.3-V supply

DEVICE POWER DISSIPATION - PD

	PARAMETER			TEST CONDITIONS	VALU E	UNITS
			0010.0	JEDEC Low-K model	231	
			SOIC-8	JEDEC High-K model	135	
θ_{JA}	Junction-to-Ambient Thermal Resistance		0010.44	JEDEC Low-K model	163	°C/W
			SOIC-14	JEDEC High-K model	92	
					73	
			SOIC-8		44	
$\theta_{\sf JB}$	Junction-to- Board Thermal Resistance		SOIC-14		61	°C/W
			QFN-20			
			SOIC-8		43	
θ_{JC}	Junction-to-Case Thermal Resistance		SOIC-14		59	°C/W
			QFN-20		14	
	Power Dissipation		HVD30,33	V _{CC} = 3.3V, T _J = 25°C,		
	Driver and receiver enabled, 50% duty	Typical	HVD31,34	$R_L = 60 \Omega$, $C_L = 50 pF (driver)$,		mW
_	cycle square-wave signal at signaling rate:		HVD32,35	C _L = 15 pF (receiver)		
P_D	HVD30,33 at 25 Mbps,		HVD30,33	$V_{CC} = 3.6V, T_{J} = 140^{\circ}C,$	197	
	HVD31,34 at 5 Mbps, HVD32,35 at 1 Mbps	Worst-case	HVD31,34	$R_L = 54 \Omega$, $C_L = 50 pF (driver)$,	213	mW
	TVD32,33 at 1 widps		HVD32,35	C _L = 15 pF (receiver)	248	
T _{SD}	Thermal Shut-down Junction Temperatur	е			170	°C



PARAMETER MEASUREMENT INFORMATION

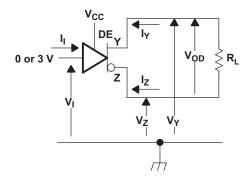


Figure 1. Driver V_{OD} Test Circuit and Voltage and Current Definitions

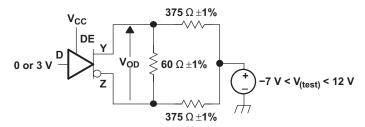


Figure 2. Driver V_{OD} With Common-Mode Loading Test Circuit

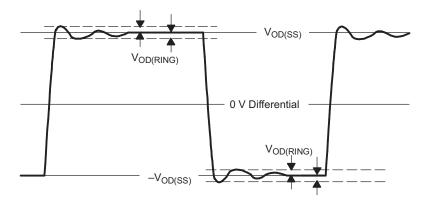
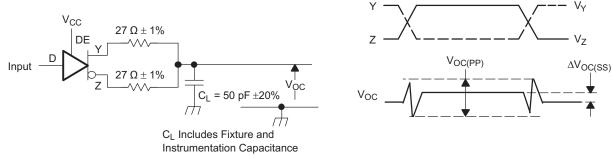


Figure 3. $V_{\text{OD(RING)}}$ Waveform and Definitions

 $V_{\text{OD(RING)}}$ is measured at four points on the output waveform, corresponding to overshoot and undershoot from the $V_{\text{OD(H)}}$ and $V_{\text{OD(L)}}$ steady state values.

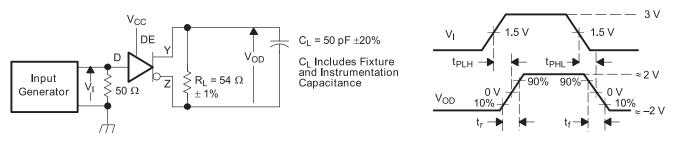


PARAMETER MEASUREMENT INFORMATION (continued)



Input: PRR = 500 kHz, 50% Duty Cycle,t $_{\rm f}$ <6ns, $t_{\rm f}$ <6ns, $Z_{\rm O}$ = 50 Ω

Figure 4. Test Circuit and Definitions for the Driver Common-Mode Output Voltage



Generator: PRR = 500 kHz, 50% Duty Cycle, t_r <6 ns, t_f <6 ns, Z_o = 50 Ω

Figure 5. Driver Switching Test Circuit and Voltage Waveforms

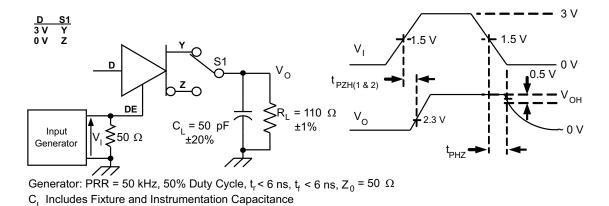
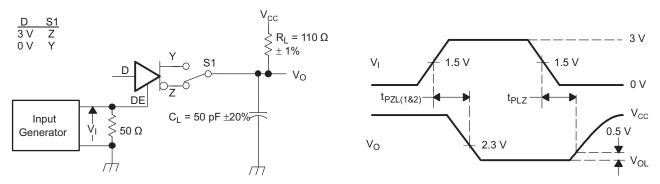


Figure 6. Driver High-Level Output Enable and Disable Time Test Circuit and Voltage Waveforms



PARAMETER MEASUREMENT INFORMATION (continued)



Generator: PRR = 50 kHz, 50% Duty Cycle, t_r < 6 ns, t_f < 6 ns, Z_0 = 50 Ω

 \boldsymbol{C}_{L} Includes Fixture and Instrumentation Capacitance

Figure 7. Driver Low-Level Output Enable and Disable Time Test Circuit and Voltage Waveforms

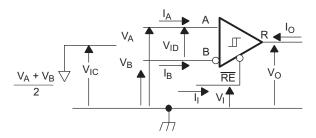


Figure 8. Receiver Voltage and Current Definitions

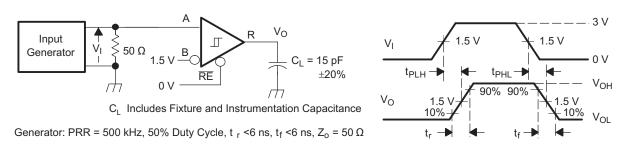
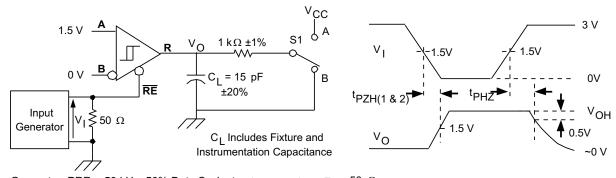


Figure 9. Receiver Switching Test Circuit and Voltage Waveforms

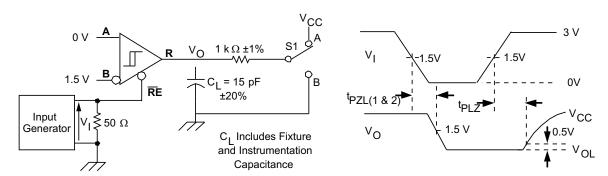


Generator: PRR = 50 kHz, 50% Duty Cycle, t_r < 6 ns, t_f < 6 ns, Z_0 = 50 Ω

Figure 10. Receiver High-Level Enable and Disable Time Test Circuit and Voltage Waveforms



PARAMETER MEASUREMENT INFORMATION (continued)



Generator: PRR = 50 kHz, 50% Duty Cycle, $\rm t_r$ < 6 ns, $\rm t_f$ < 6 ns, $\rm Z_0$ = 50 $\rm \Omega$

Figure 11. Receiver Enable Time From Standby (Driver Disabled)

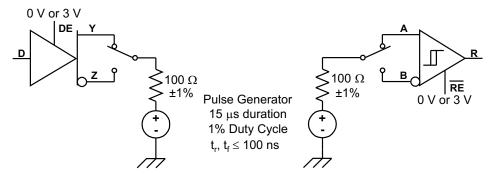


Figure 12. Test Circuit, Transient Over Voltage Test



DEVICE INFORMATION

LOW-POWER STANDBY MODE

When both the driver and receiver are disabled (DE low and $\overline{\text{RE}}$ high) the device is in standby mode. If the enable inputs are in this state for less than 60 ns, the device does not enter standby mode. This guards against inadvertently entering standby mode during driver/receiver enabling. Only when the enable inputs are held in this state for 300 ns or more, the device is assured to be in standby mode. In this low-power standby mode, most internal circuitry is powered down, and the supply current is typically less than 1 nA. When either the driver or the receiver is re-enabled, the internal circuitry becomes active.

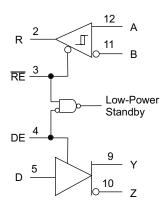


Figure 13. Low-Power Standby Logic Diagram

If only the driver is re-enabled (DE transitions to high) the driver outputs are driven according to the D input after the enable times given by t_{PZH2} and t_{PZL2} in the driver switching characteristics. If the D input is open when the driver is enabled, the driver outputs defaults to A high and B low, in accordance with the driver failsafe feature.

If only the receiver is re-enabled (\overline{RE} transitions to low) the receiver output is driven according to the state of the bus inputs (A and B) after the enable times given by t_{PZH2} and t_{PZL2} in the receiver switching characteristics. If there is no valid state on the bus the receiver responds as described in the failsafe operation section.

If both the receiver and driver are re-enabled simultaneously, the receiver output is driven according to the state of the bus inputs (A and B) and the driver output is driven according to the D input. Note that the state of the active driver affects the inputs to the receiver. Therefore, the receiver outputs are valid as soon as the driver outputs are valid.

DRIVER OUTPUT CURRENT LIMITING

The RS-485 standard (ANSI/TIA/EIA-485-A or equivalently ISO 8482) specifies a 250 mA driver output current limit to prevent damage caused by data contention on the bus. That applies in the event that two or more transceivers drive the bus to opposing states at the same time. The HVD3x family of devices includes current limiting circuitry that prevents damage under these conditions. Note that this current limit prevents damage during the bus contention, but the logic state of the bus may be indeterminate as specified by the standard, so communication errors may occur.

In a specific combination of circumstances, a condition may occur in which current through the bus pin exceeds the 250 mA limit. This combination of conditions is not normally included in RS-485 applications:

- loading capacitance on the pin is less than 500 pF
- the bus pin is directly connected to a voltage more negative than -1V
- the device is supplied with Vcc equal or greater than 3.3V
- the driver is enabled
- the bus pin is driving to the logic high state.

In these specific conditions, the normal current limit circuitry and thermal shutdown circuitry will not limit or shutdown the current flow. If the current is allowed to continue, the device will heat up in a localized area near the driver outputs, and the device may be damaged.

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Typical RS-485 twisted-pair cable has capacitance of approximately 50 pF/meter. Therefore it is expected that 10 meters of cable would provide sufficient capacitance to prevent this latch-up condition.

The -7 to +12V common mode range specified by RS-485 is intended to allow communication between transceivers separated by significant distances, when ground offsets may occur due to temporary current surges, electrical noise, etc. In those circumstances, the inherent cable needed to connect separated transceivers will ensure that the conditions above do not occur. For transceiver separated by only a short cable length, or backplane applications, it would be unusual for there to be a steady-state negative common-mode voltage. It is possible for a negative power supply to be shorted to the bus lines due to mis-wiring or cable damage, however, this is a different root cause fault, and robust devices such as the HVD178x family should be used for surviving power supply or mis-wiring faults.

The 250 mA current limit in the RS-485 standard is intended to prevent damage caused by data contention on the bus; that is, in the event that two or more transceivers drive the bus to different states at the same time. These devices will not be damaged under these conditions, because all RS-485 drivers have output impedance sufficient to prevent the direct connection condition stated above. Typical RS-485 driver output impedance is on the order of 10 to $30~\Omega$.

HOT-PLUGGING

These devices are designed to operate in *hot swap* or *hot pluggable* applications. Key features for hot-pluggable applications are power-up, power-down glitch-free operation, default disabled input/output pins, and receiver failsafe. As shown in Figure 24, an internal Power-On Reset circuit keeps the driver outputs in a high-impedance state until the supply voltage has reached a level at which the device will reliably operate. This ensures that no spurious bits are transmitted on the bus pin outputs as the power supply turns on or turns off.

As shown in the device FUNCTION TABLES, the enable inputs have the feature of default disable on both the driver enable and receiver enable. This ensures that the device will neither drive the bus nor report data on the R pin until the associated controller actively drives the enable pins.

RECEIVER FAILSAFE

The differential receiver is *failsafe* to invalid bus states caused by open bus conditions such as, a disconnected connector, shorted bus conditions caused by damaged cabling, or idle bus conditions that occur when no driver is actively driving a valid RD-485 bus state on the network. In any of these cases, the differential receiver will output a failsafe HIGH state, so that small noise signals do not cause spurious transitions at the receiver output.

SAFE OPERATION WITH BUS CONTENTION

These devices incorporate a driver current limit of 250 mA across the RS-485 common-mode range of -7 V to +12 V. As stated in the "Application Guidelines for TIA/EIA-485-A" (1) this sets a practical limitation to prevent damage during bus contention events. Contention can occur during system initialization, during system faults, or whenever two or more drivers are active at the same time.

Figure 14 shows a 2-node system to demonstrate bus contention by forcing both drivers to be active in opposing states.



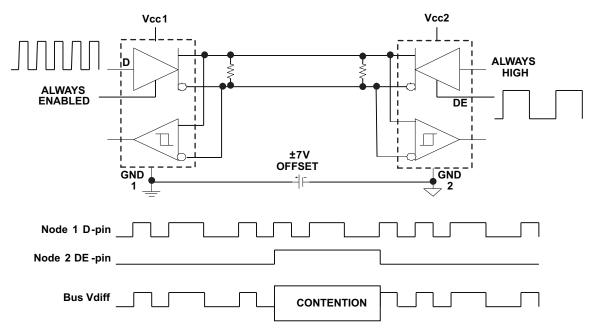


Figure 14. Bus Contention Example

Figure 15 shows typical operation in a bus contention event. The bottom trace illustrates how the SN65HVD33 at Node 1 continues normal operation after a contention event between the two drivers, with a –7 V ground offset on Node 2. This illustrates how the HVD3x family of devices operates robustly in spite of bus contention faults, even with large common-mode offsets.

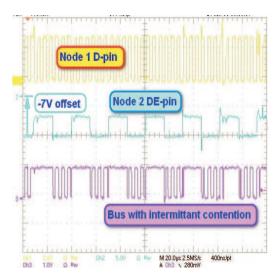


Figure 15. HVD3x Drivers Operate Correctly After Bus Contention Faults



FUNCTION TABLES

Table 1. SN65HVD33, SN65HVD34, SN65HVD35 DRIVER

IN	PUTS	OUTPUTS			
D	DE	Y	Z		
Н	Н	Н	L		
L	Н	L	Н		
Х	L or open	Z	Z		
Open	Н	L	Н		

Table 2. SN65HVD33, SN65HVD34, SN65HVD35 RECEIVER

DIFFERENTIAL INPUTS $V_{ID} = V_{(A)} - V_{(B)}$	ENABLE RE	OUTPUT R
V _{ID} ≤ -0.2 V	L	L
-0.2 V < V _{ID} < -0.02 V	L	?
-0.02 V ≤ V _{ID}	L	П
X	H or open	Z
Open Circuit	L	Н
Idle circuit	L	Н
Short Circuit, $V_{(A)} = V_{(B)}$	L	Н

Table 3. SN65HVD30, SN65HVD31, SN65HVD32 DRIVER

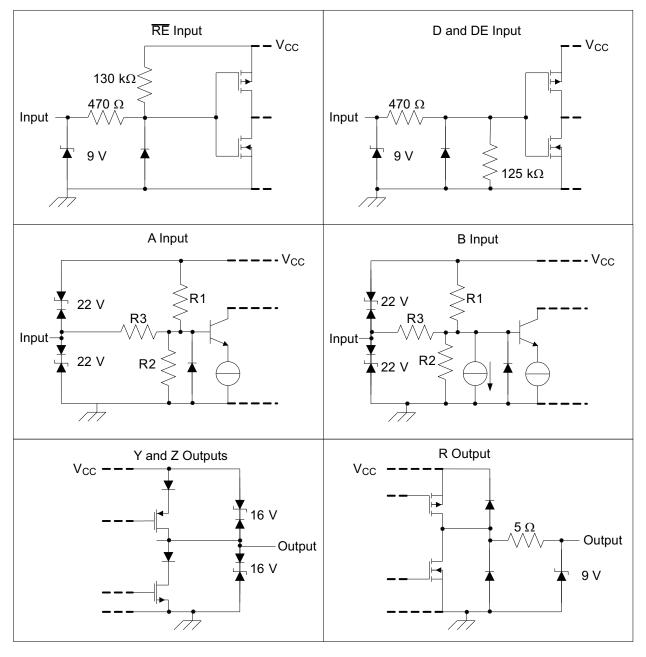
	OUTPUTS					
INPUT D	Y	Z				
Н	Н	L				
L	L	Н				
Open	L	Н				

Table 4. SN65HVD30, SN65HVD31, SN65HVD32 RECEIVER

DIFFERENTIAL INPUTS $V_{ID} = V_{(A)} - V_{(B)}$	OUTPUT R
V _{ID} ≤ −0.2 V	L
-0.2 V < V _{ID} < -0.02 V	?
-0.02 V ≤ V _{ID}	Н
Open Circuit	Н
Idle circuit	Н
Short Circuit, V _(A) = V _(B)	Н



EQUIVALENT INPUT AND OUTPUT SCHEMATIC DIAGRAMS



	R1/R2	R3
SN65HVD30, SN65HVD33	9 kΩ	45 kΩ
SN65HVD31, SN65HVD32, SN65HVD34, SN65HVD35	36 kΩ	180 kΩ

30

0

5



TYPICAL CHARACTERISTICS

RMS SUPPLY CURRENT VS SIGNALING RATE TA = 25°C RL = 54 \Omega RE = V_{CC} CL = 50 pF DE = V_{CC} TO DE = V_{CC} 35 35

10

15

Signaling Rate - Mbps Figure 16.

20

HVD30, HVD33

HVD31, HVD34 RMS SUPPLY CURRENT VS SIGNALING RATE

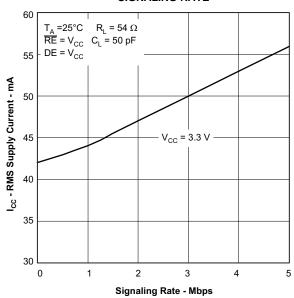
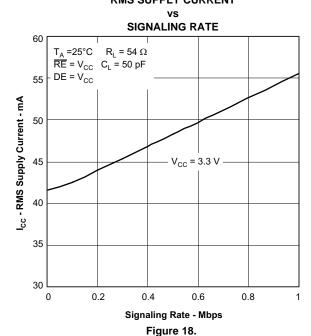


Figure 17.

HVD32, HVD35 RMS SUPPLY CURRENT

25



HVD31, HVD32, HVD34, HVD35



TYPICAL CHARACTERISTICS (continued)

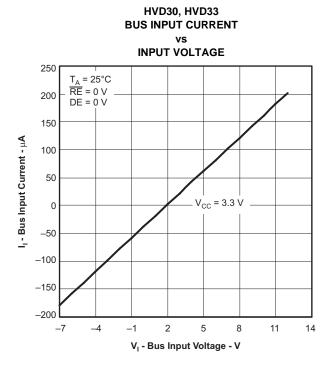


Figure 19.

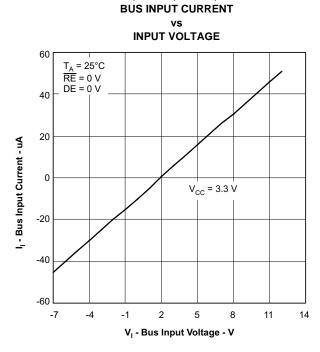


Figure 20.

DRIVER DIFFERENTIAL OUTPUT VOLTAGE

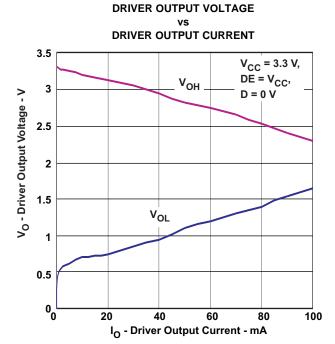


Figure 21.

vs **DRIVER OUTPUT CURRENT** 3.5 $V_{CC} = 3.3 \text{ V},$ DE = V_{CC} , **100** Ω **60** Ω Vo - Driver Differential Output Voltage - V 3 D = 0 V2.5 2 1.5 1 0.5 0 0 20 40 60 80 100 IO - Driver Output Current - mA

Figure 22.

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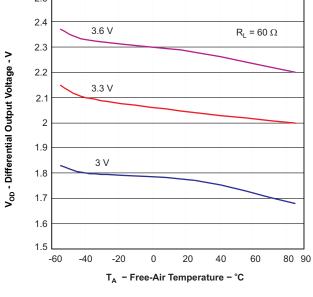


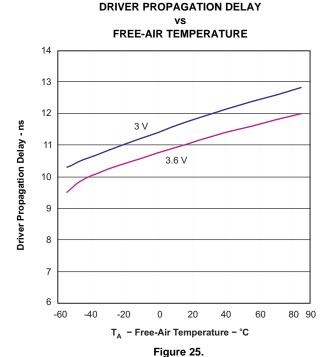
TYPICAL CHARACTERISTICS (continued)

DRIVER DIFFERENTIAL OUTPUT VOLTAGE VS FREE-AIR TEMPERATURE 2.5 2.4 3.6 V $R_I = 60 \Omega$ 2.3 2.2 3.3 V 2.1 2 1.9 3 V 1.8

Figure 23.

HVD30, HVD33





DRIVER OUTPUT CURRENT

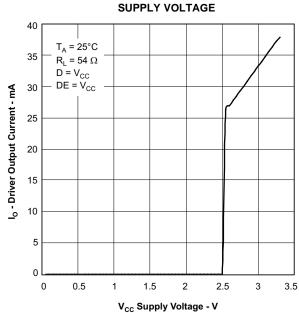


Figure 24.



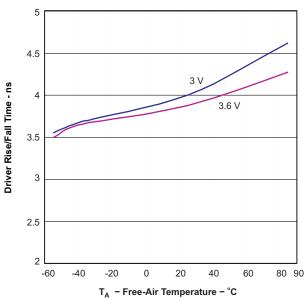
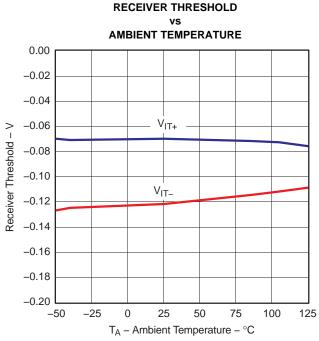


Figure 26.



TYPICAL CHARACTERISTICS (continued)





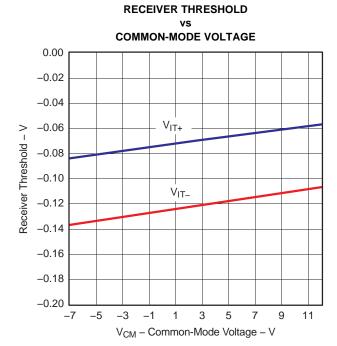


Figure 28.

ENABLE TIME

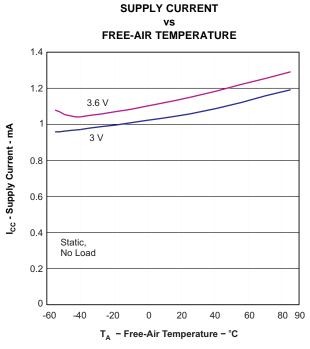


Figure 29.

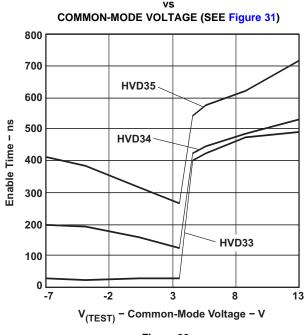


Figure 30.

TYPICAL CHARACTERISTICS (continued)

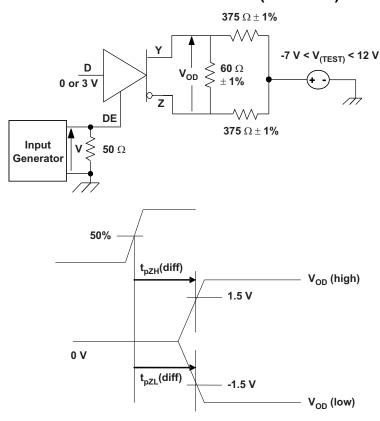


Figure 31. Driver Enable Time From DE to $V_{\rm OD}$

The time $t_{pZL}(x)$ is the measure from DE to $V_{OD}(x)$. V_{OD} is valid when it is greater than 1.5 V.

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
SN65HVD30D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD30DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD30DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD30DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD31D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD31DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD31DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD31DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD32D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD32DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD32DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD32DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD33D	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD33DG4	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD33DR	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD33DRG4	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD33RHLR	ACTIVE	QFN	RHL	20	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	





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Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
SN65HVD33RHLT	ACTIVE	QFN	RHL	20	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
SN65HVD34D	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD34DG4	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD34DR	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD34DRG4	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD35D	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD35DG4	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD35DR	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
SN65HVD35DRG4	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.



PACKAGE OPTION ADDENDUM

19-Oct-2011

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF SN65HVD30, SN65HVD33:

Enhanced Product: SN65HVD30-EP, SN65HVD33-EP

NOTE: Qualified Version Definitions:

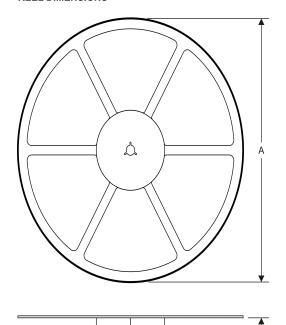
• Enhanced Product - Supports Defense, Aerospace and Medical Applications

PACKAGE MATERIALS INFORMATION

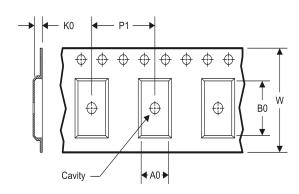
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TAPE AND REEL INFORMATION

REEL DIMENSIONS



TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
SN65HVD30DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD31DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD32DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD33DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
SN65HVD33RHLR	QFN	RHL	20	3000	330.0	12.4	3.8	4.8	1.6	8.0	12.0	Q1
SN65HVD33RHLT	QFN	RHL	20	250	180.0	12.4	3.8	4.8	1.6	8.0	12.0	Q1
SN65HVD34DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
SN65HVD35DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1

PACKAGE MATERIALS INFORMATION

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
SN65HVD30DR	SOIC	D	8	2500	367.0	367.0	35.0
SN65HVD31DR	SOIC	D	8	2500	367.0	367.0	35.0
SN65HVD32DR	SOIC	D	8	2500	367.0	367.0	35.0
SN65HVD33DR	SOIC	D	14	2500	367.0	367.0	38.0
SN65HVD33RHLR	QFN	RHL	20	3000	367.0	367.0	35.0
SN65HVD33RHLT	QFN	RHL	20	250	210.0	185.0	35.0
SN65HVD34DR	SOIC	D	14	2500	367.0	367.0	38.0
SN65HVD35DR	SOIC	D	14	2500	367.0	367.0	38.0

D (R-PDSO-G14)

PLASTIC SMALL OUTLINE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AB.



D (R-PDSO-G14)

PLASTIC SMALL OUTLINE

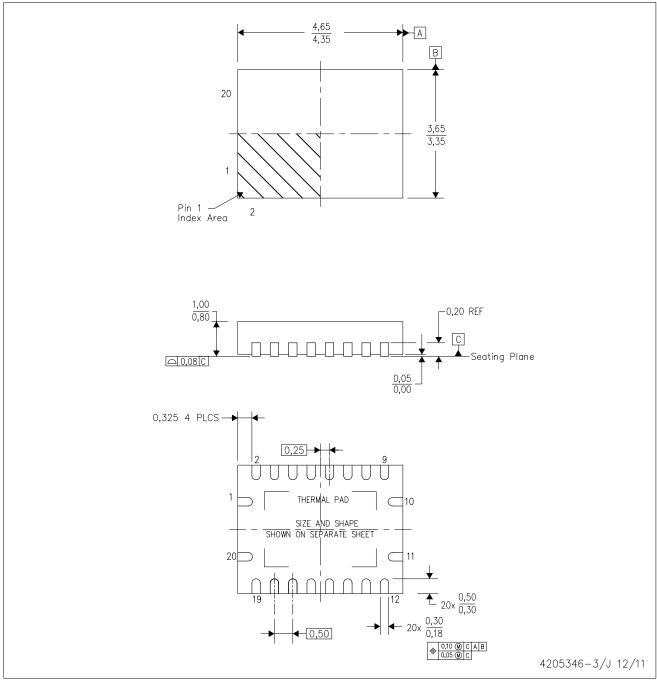


- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



RHL (R-PVQFN-N20)

PLASTIC QUAD FLATPACK NO-LEAD



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.

- B. This drawing is subject to change without notice.
- C. QFN (Quad Flatpack No-Lead) Package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.



RHL (S-PVQFN-N20)

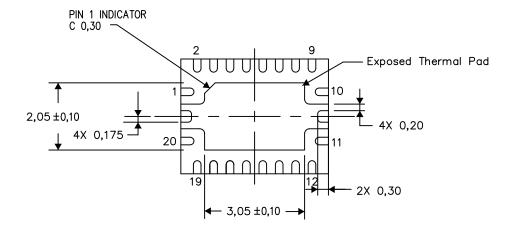
PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



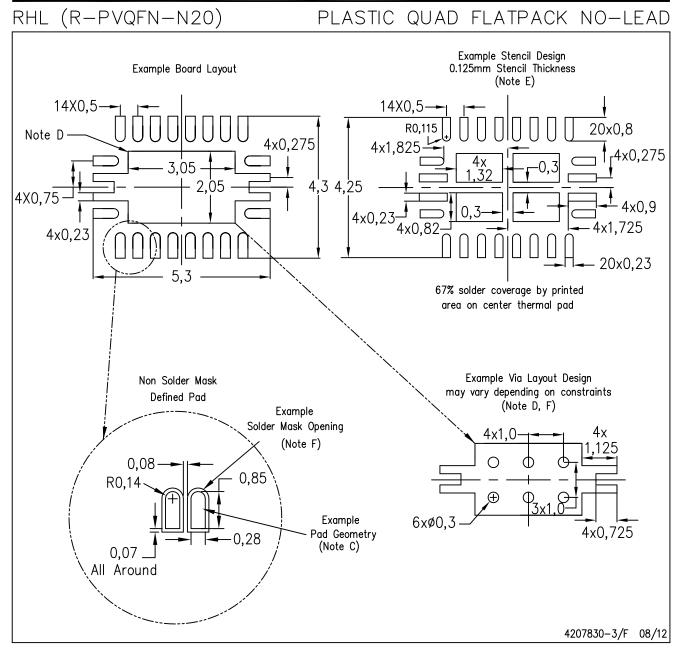
Bottom View

Exposed Thermal Pad Dimensions

4206363-3/M 08/12

NOTE: All linear dimensions are in millimeters





- : A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Publication IPC-7351 is recommended for alternate designs.
 - D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat—Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com http://www.ti.com.
 - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
 - F. Customers should contact their board fabrication site for minimum solder mask web tolerances between signal pads.



D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.



D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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