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## 8-PORT LVDS REPEATER

#### **FEATURES**

- One Line Receiver and Eight Line Drivers Configured as an 8-Port LVDS Repeater
- Line Receiver and Line Drivers Meet or Exceed the Requirements of ANSI EIA/TIA-644 Standard
- Typical Data Signaling Rates to 400 Mbps or **Clock Frequencies to 400 MHz**
- **Enabling Logic Allows Individual Control of Each Driver Output, Plus All Outputs**
- **Low-Voltage Differential Signaling With** Typical Output Voltage of 350 mV and a 100- $\Omega$
- **Electrically Compatible With LVDS, PECL,** LVPECL, LVTTL, LVCMOS, GTL, BTL, CTT, SSTL, or HSTL Outputs With External **Termination Networks**
- Propagation Delay Times < 4.7 ns
- Output Skew Less Than 300 ps and Part-to-Part Skew Less Than 1.5 ns
- **Total Power Dissipation at 200 MHz Typically** Less Than 330 mW With 8 Channels Enabled
- **Driver Outputs or Receiver Input Equals High** Impedance When Disabled or With  $V_{CC}$  < 1.5 V
- **Bus-Pin ESD Protection Exceeds 12 kV**
- Packaged in Thin Shrink Small-Outline Package With 20-Mil Terminal Pitch

#### **DBT PACKAGE** (TOP VIEW) GND d 1 ○ **NC** 38 V<sub>CC</sub> [] 2 AY 37 GND ∏3 ΠAZ 36 NC ∏ 4 BY 35 ENM ∏ 5 BZ 34 ENA ∏ 6 ПCY 33 ENB **∏** 7 ηcz 32 ENC ∏8 DY 31 END 9 **D**DZ 30 A [] 10 ΠEΥ 29 B 🛮 11 ΠEZ 28 27 | FY ENE ∏ 12 ENF **1**3 **∏** FZ 26 ENG ∏ 14 **∏**GY 25 ENH **1** 15 **∏** GZ 24 NC **∏** 16 23 HY GND [] 17 22 | HZ ] NC V<sub>CC</sub> 18 21 **∏**NC GND [] 19 20

NC - No internal connection

### DESCRIPTION

The SN65LVDS108 is configured as one differential line receiver connected to eight differential line drivers. Individual output enables are provided for each output and an additional enable is provided for all outputs.

The line receivers and line drivers implement the electrical characteristics of low-voltage differential signaling (LVDS). LVDS, as specified in EIA/TIA-644, is a data signaling technique that offers low power, low noise emission, high noise immunity, and high switching speeds. (Note: The ultimate rate and distance of data transfer is dependent upon the attenuation characteristics of the media, the noise coupling to the environment, and other system characteristics.)

The intended application of this device, and the LVDS signaling technique, is for point-to-point or point-to-multipoint (distributed simplex) baseband data transmission on controlled impedance media of approximately 100  $\Omega$ . The transmission media may be printed-circuit board traces, backplanes, or cables. The large number of drivers integrated into the same silicon substrate, along with the low pulse skew of balanced signaling, provides extremely precise timing alignment of the signals being repeated from the inputs. This is particularly advantageous for implementing system clock or data distribution trees.

The SN65LVDS108 is characterized for operation from -40°C to 85°C.



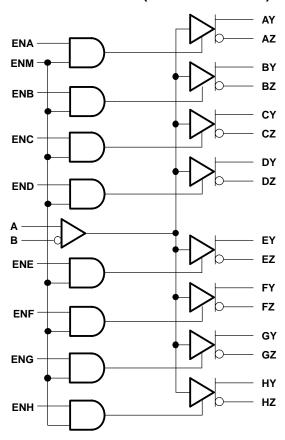
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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## **LOGIC DIAGRAM (POSITIVE LOGIC)**



## **SELECTION GUIDE TO LVDS SPLITTER**

The SN65LVDS108 is one member of a family of LVDS splitters and repeaters. A brief overview of the family is provided in the following table.

## LVDS SPLITTER AND REPEATER FAMILY

DEVICE	NUMBER OF INPUTS	NUMBER OF OUTPUTS	PACKAGE	COMMENTS
SN65LVDS104	1 LVDS	4 LVDS	16-pin D	4-Port LVDS repeater
SN65LVDS105	1 LVTTL	4 LVDS	16-pin D	4-Port TTL-to-LVDS repeater
SN65LVDS108	1 LVDS	8 LVDS	38-pin DBT	8-Port LVDS repeater
SN65LVDS109	2 LVDS	8 LVDS	38-pin DBT	Dual 4-port LVDS repeater
SN65LVDS116	1 LVDS	16 LVDS	64-pin DGG	16-Port LVDS repeater
SN65LVDS117	2 LVDS	16 LVDS	64-pin DGG	Dual 8-port LVDS repeater

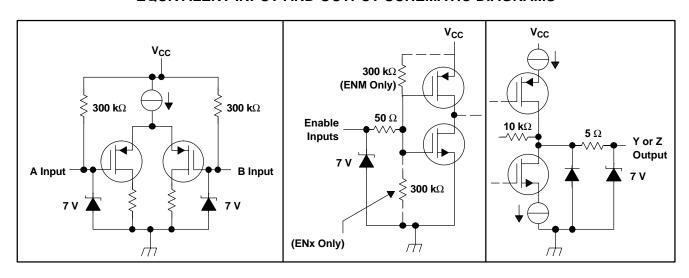


## **FUNCTION TABLE**(1)

INPUTS	INPUTS			
$V_{ID} = V_A - V_B$	ENM	ENx	xΥ	χZ
X	L	Х	Z	Z
X	Х	L	Z	Z
V <sub>ID</sub> ≥ 100 mV	Н	Н	Н	L
-100 mV < V <sub>ID</sub> < 100 mV	Н	Н	?	?
V <sub>ID</sub> ≤-100 mV	Н	Н	L	Н

(1) H = high level, L = low level, Z = high impedance, X = don't care, ? = indeterminate

## **EQUIVALENT INPUT AND OUTPUT SCHEMATIC DIAGRAMS**



#### **ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature range (unless otherwise noted)(1)

			UNIT
$V_{CC}$	Supply voltage range <sup>(2)</sup>		-0.5 V to 4 V
	Input voltage range	Enable inputs	-0.5 V to 6 V
	Input voltage range	A, B, Y or Z	-0.5 V to 4 V
	Electrostatic discharge, A, E	3, Y, Z, and GND <sup>(3)</sup>	Class 3, A:12 kV, B: 500 V
	Continuous power dissipation	on	See Dissipation Rating Table
	Storage temperature range		-65°C to 150°C
	Lead temperature 1,6 mm (	260°C	

<sup>(1)</sup> Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

## **DISSIPATION RATING TABLE**

PACKAGE	$T_A \le 25^{\circ}C$ POWER RATING	DERATING FACTOR <sup>(1)</sup> ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 85°C POWER RATING
DBT	1277 mW	10.2 mW/°C	644 mW

<sup>(1)</sup> This is the inverse of the junction-to-ambient thermal resistance when board-mounted (low-k) with no air flow.

<sup>(2)</sup> All voltage values, except differential I/O bus voltages, are with respect to network ground terminal.

<sup>3)</sup> Tested in accordance with MIL-STD-883C Method 3015.7.



## RECOMMENDED OPERATING CONDITIONS

		MIN	NOM	MAX	UNIT
V <sub>CC</sub>	Supply voltage	3	3.3	3.6	V
V <sub>IH</sub>	High-level input voltage	2			V
V <sub>IL</sub>	Low-level input voltage			0.8	V
V <sub>I</sub> or V <sub>IC</sub>	Voltage at any bus terminal (separately or common-mode)	0		V <sub>CC</sub> - 0.8	V
T <sub>A</sub>	Operating free-air temperature	40		85	°C

## **ELECTRICAL CHARACTERISTICS**

over recommended operating conditions (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX	UNIT
$V_{\text{ITH+}}$	Positive-going differential input voltage threshold	See Figure 1 and Table 1			100	mV
$V_{\text{ITH-}}$	Negative-going differential input voltage threshold	See Figure 1 and Table 1	100			IIIV
V <sub>OD</sub>	Differential output voltage magnitude	D 400 O V 1400 mV Coo	247	340	454	
$\Delta  V_{OD} $	Change in differential output voltage magnitude between logic states	$R_L$ = 100 Ω, $V_{ID}$ = ±100 mV, See Figure 1 and Figure 2	50		50	mV
V <sub>OC(SS</sub>	Steady-state common-mode output voltage		1.125		1.375	>
$\Delta V_{OC(}$	Change in steady-state common-mode output voltage between logic states	See Figure 3	50		50	mV
V <sub>OC(PP</sub> )	Peak-to-peak common-mode output voltage			50	150	IIIV
	Supply current	Enabled, $R_L = 100 \Omega$		62	85	mA
I <sub>CC</sub>	Supply current	Disabled		8	12	ША
	Input current (A or B inputs)	$V_I = 0 V$	2		20	
I <sub>I</sub>	input current (A or B inputs)	$V_1 = 2.4 \text{ V}$	1.2			μΑ
I <sub>I(OFF)</sub>	Power-off input current (A or B inputs)	$V_{CC} = 1.5 \text{ V}, \qquad V_{I} = 2.4 \text{ V}$			20	μΑ
I <sub>IH</sub>	High-level input current (enables)	V <sub>IH</sub> = 2 V			±20	μΑ
I <sub>IL</sub>	Low-level input current (enables)	V <sub>IL</sub> = 0.8 V			±10	μΑ
	Chart aircuit autaut aurrant	V <sub>OY</sub> or V <sub>OZ</sub> = 0 V			±24	A
los	Short-circuit output current	V <sub>OD</sub> = 0 V			±12	mA
I <sub>OZ</sub>	High-impedance output current	$V_O = 0 \text{ V or } V_{CC}$			±1	μΑ
I <sub>O(OFF)</sub>	Power-off output current	$V_{CC} = 1.5 \text{ V}, \qquad V_{O} = 3.6 \text{ V}$			±1	μΑ
C <sub>IN</sub>	Input capacitance (A or B inputs)	$V_I = 0.4 \sin (4E6\pi t) + 0.5 V$		5		
C <sub>O</sub>	Output capacitance (Y or Z outputs)	$V_1 = 0.4 \sin (4E6\pi t) + 0.5 V$ , Disabled		9.4		pF

<sup>(1)</sup> All typical values are at  $25^{\circ}$ C and with a 3.3-V supply.



# **SWITCHING CHARACTERISTICS**

over recommended operating conditions (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output		1.6	2.8	4.5	20
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output		1.6	2.8	4.5	ns
t <sub>r</sub>	Differential output signal rise time		0.3	0.8	1.2	20
t <sub>f</sub>	Differential output signal fall time	$R_L = 100 \Omega$ , $C_L = 10$ pF, See Figure 4	0.3	0.8	1.2	ns
t <sub>sk(p)</sub>	Pulse skew ( t <sub>PHL</sub> - t <sub>PLH</sub>  ) <sup>(2)</sup>	pr, occ riguio 4		150	500	20
t <sub>sk(o)</sub>	Output skew <sup>(3)</sup>				300	ps
t <sub>sk(pp)</sub>	Part-to-part skew <sup>(4)</sup>				1.5	ns
t <sub>PZH</sub>	Propagation delay time, high-impedance-to-high-level output			5.7	15	
t <sub>PZL</sub>	Propagation delay time, high-impedance-to-low-level output	Coo Figure F		7.7	15	ns
t <sub>PHZ</sub>	Propagation delay time, high-level-to-high-impedance output	See Figure 5		3.2	15	
t <sub>PLZ</sub>	Propagation delay time, low-level-to-high-impedance output			3.2	15	ns

- All typical values are at 25°C and with a 3.3-V supply.
- (3) (4)
- $t_{sk(p)}$  is the magnitude of the time difference between the  $t_{PLH}$  and  $t_{PHL}$  of any output of a single device.  $t_{sk(p)}$  is the magnitude of the time difference between the  $t_{PLH}$  or  $t_{PHL}$  measured at any two outputs.  $t_{sk(pp)}$  is the magnitude of the time difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.

## PARAMETER MEASUREMENT INFORMATION

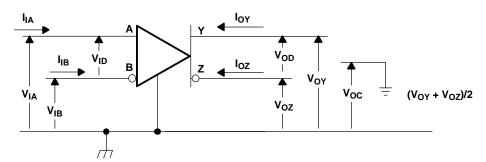


Figure 1. Voltage and Current Definitions

Table 1. Receiver Minimum and Maximum Input Threshold Test Voltages

APPLIED	VOLTAGES	RESULTING DIFFERENTIAL INPUT VOLTAGE	RESULTING COMMON- MODE INPUT VOLTAGE
V <sub>IA</sub>	V <sub>IB</sub>	V <sub>ID</sub>	V <sub>IC</sub>
1.25 V	1.15 V	100 mV	1.2 V
1.15 V	1.25 V	-100 mV	1.2 V
2.4 V	2.3 V	100 mV	2.35 V
2.3 V	2.4 V	–100 mV	2.35 V
0.1 V	0 V	100 mV	0.05 V
0 V	0.1 V	–100 mV	0.05 V
1.5 V	0.9 V	600 mV	1.2 V
0.9 V	1.5 V	−600 mV	1.2 V
2.4 V	1.8 V	600 mV	2.1 V
1.8 V	2.4 V	−600 mV	2.1 V
0.6 V	0 V	600 mV	0.3 V
0 V	0.6 V	−600 mV	0.3 V



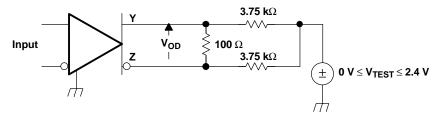
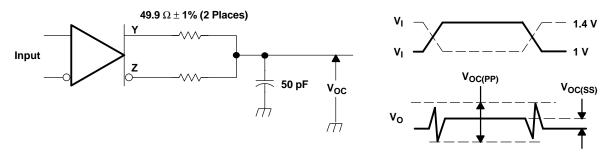
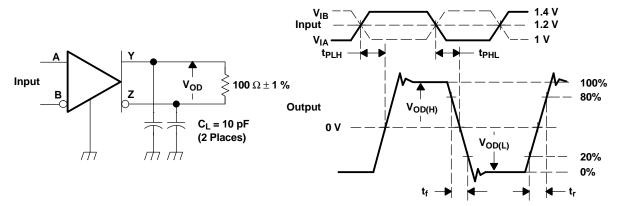


Figure 2. V<sub>OD</sub> Test Circuit



A. All input pulses are supplied by a generator having the following characteristics: t<sub>r</sub> or t<sub>f</sub> ≤ 1 ns, pulse repetition rate (PRR) = 0.5 Mpps, pulsewidth = 500 ±10 ns. C<sub>L</sub> includes instrumentation and fixture capacitance within 0,06 mm of the D.U.T. The measurement of V<sub>OC(PP)</sub> is made on test equipment with a −3 dB bandwidth of at least 300 MHz.

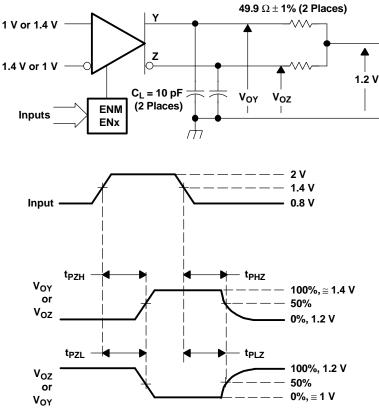
Figure 3. Test Circuit and Definitions for the Driver Common-Mode Output Voltage



A. All input pulses are supplied by a generator having the following characteristics:  $t_r$  or  $t_r \le 1$  ns, pulse repetition rate (PRR) = 50 Mpps, pulsewidth = 10  $\pm 0.2$  ns .  $C_L$  includes instrumentation and fixture capacitance within 0,06 mm of the D.U.T.

Figure 4. Test Circuit, Timing, and Voltage Definitions for the Differential Output Signal



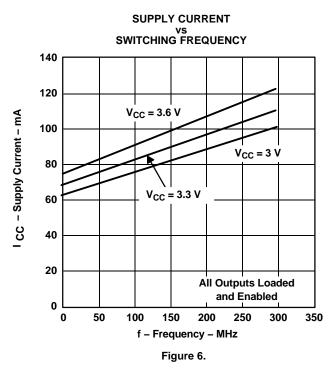


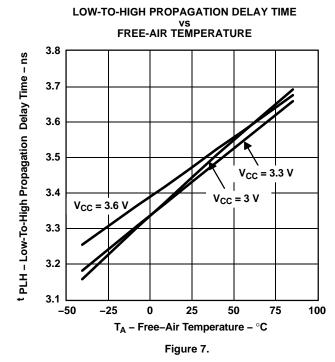
A. All input pulses are supplied by a generator having the following characteristics:  $t_r$  or  $t_f \le 1$  ns, pulse repetition rate (PRR) = 0.5 Mpps, pulsewidth = 500  $\pm 10$  ns .  $C_L$  includes instrumentation and fixture capacitance within 0,06 mm of the D.U.T.

Figure 5. Enable and Disable Time Circuit and Definitions

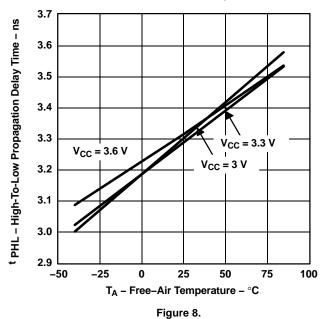


## **TYPICAL CHARACTERISTICS**





# HIGH-TO-LOW PROPAGATION DELAY TIME vs FREE-AIR TEMPERATURE



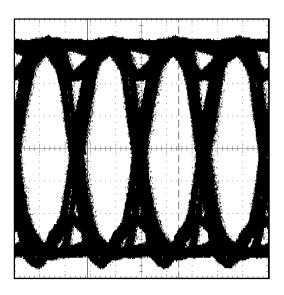
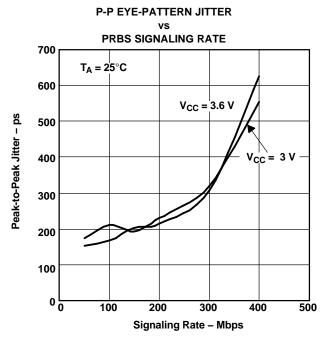


Figure 9. Typical Differential Eye Pattern at 400 Mbps

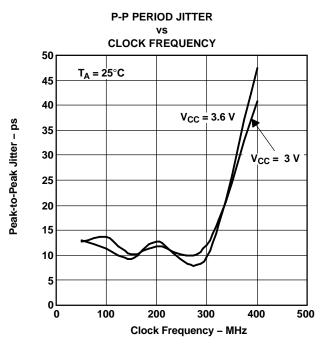


## **TYPICAL CHARACTERISTICS (continued)**



NOTES: Input:  $2^{15}$  PRBS with peak-to-peak jitter < 100 ps at 100 Mbps, all outputs enabled and loaded with differential  $100-\Omega$  loads, worst-case output, supply decoupled with  $0.1-\mu F$  and  $0.001-\mu F$  ceramic 0603-style capacitors 1 cm from the device.

Figure 10.



NOTES: Input: 50% duty cycle square wave with period jitter < 9 ps at 100 MHz, all outputs enabled and loaded with differential  $100-\Omega$  loads,worst-case output, supply decoupled with 0.1- $\mu$ F and 0.001-  $\mu$ F ceramic 0603-style capacitors 1 cm from the device.

Figure 11.

#### **APPLICATION INFORMATION**

#### **FAIL SAFE**

A common problem with differential signaling applications is how the system responds when no differential voltage is present on the signal pair. The LVDS receiver is like most differential line receivers, in that its output logic state can be indeterminate when the differential input voltage is between –100 mV and 100 mV and within its recommended input common-mode voltage range. Hovever, TI LVDS receivers handles the open-input circuit situation differently.

Open-circuit means that there is little or no input current to the receiver from the data line itself. This could be when the driver is in a high-impedance state or the cable is disconnected. When this occurs, the LVDS receiver pulls each line of the signal pair to near  $V_{CC}$  through 300-k $\Omega$  resistors as shown in Figure 12. The fail-safe feature uses an AND gate with input voltage thresholds at about 2.3 V to detect this condition and force the output to a high-level regardless of the differential input voltage.

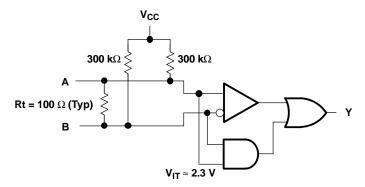


Figure 12. Open-Circuit Fail Safe of the LVDS Receiver

It is only under these conditions that the output of the receiver will be valid with less than a 100 mV differential input voltage magnitude. The presence of the termination resistor, Rt, does not affect the fail-safe function as long as it is connected as shown in Figure 12. Other termination circuits may allow a dc current to ground that could defeat the pullup currents from the receiver and the fail-safe feature.

## **CLOCK DISTRIBUTION**

The SN65LVDS108 device solves several problems common to the distribution of timing critical clock and data signals. These problems include:

- Excessive skew between the signal paths
- Noise pickup over long signaling paths
- High power consumption
- Control of which signal paths are enabled or disabled
- Elimination of radiation from unterminated lines

Buffering and splitting the signal on the same silicon die minimizes corruption of the timing relation between the copies of the signal. Buffering and splitting the signal in separate devices will introduce considerably higher levels of uncontrolled timing skew between the signals. Higher speed operation and more timing tolerance for other components of the system is enabled by the tighter system timing budgets provided by the single die implementations of the SN65LVDS108.

The use of LVDS signaling technology for both the inputs and the outputs provides superior common-mode and noise tolerance compared to single-ended I/O technologies. This is particularly important because the signals that are being distributed must be transmitted over longer distances, and at higher rates, than can be accommodated with single-ended I/Os. In addition, LVDS consumes considerably less power than other high-performance differential signaling schemes.



The enable inputs provided for each output may be used to turn on or off any of the paths. This function is required to prevent radiation of signals from the unterminated signal lines on open connectors when boards or devices are being swapped in the end equipment. The individual channel enables are also required if redundant paths are being utilized for reliability reasons.

The following diagram shows how an input signal is being identically repeated out two of the available outputs. A third output is shown in the disabled state.

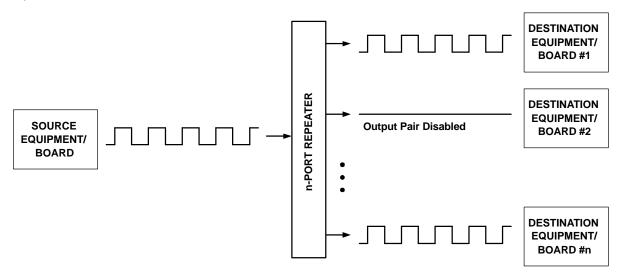


Figure 13. LVDS Repeating Splitter Application Example Showing Individual Path Control

## **INPUT LEVEL TRANSLATION**

An LVDS receiver can be used to receive various other types of logic signals. Figure 14 through Figure 22 show the termination circuits for SSTL, HSTL, CTT, GTL, BTL, LVPECL, PECL, CMOS, and TTL.

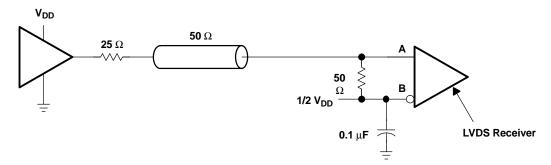


Figure 14. Stub-Series Terminated (SSTL) or High-Speed Transceiver Logic (HSTL)



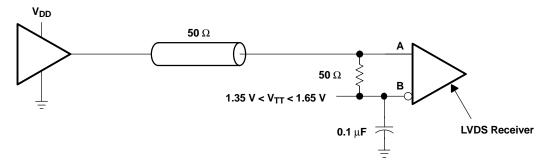


Figure 15. Center-Tap Termination (CTT)

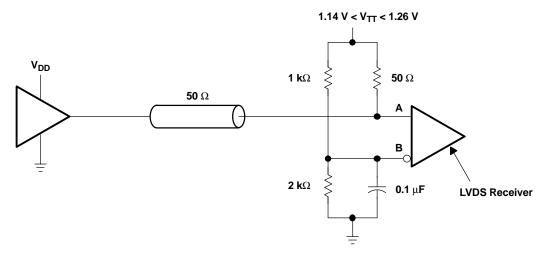


Figure 16. Gunning Transceiver Logic (GTL)

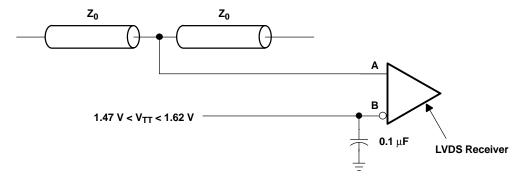


Figure 17. Backplane Transceiver Logic (BTL)



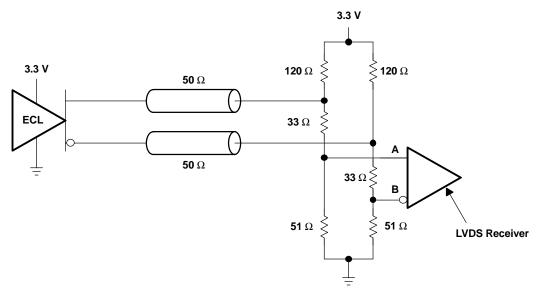


Figure 18. Low-Voltage Positive Emitter-Coupled Logic (LVPECL)

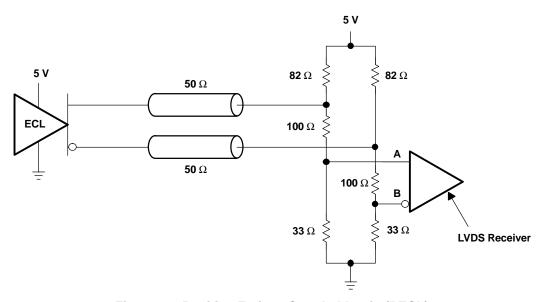


Figure 19. Positive Emitter-Coupled Logic (PECL)



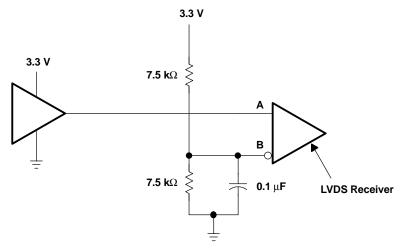


Figure 20. 3.3-V CMOS

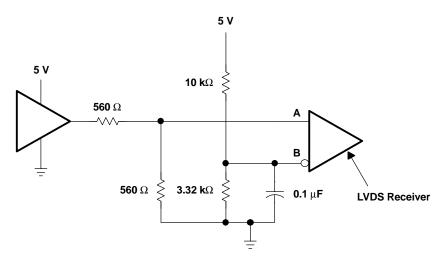


Figure 21. 5-V CMOS

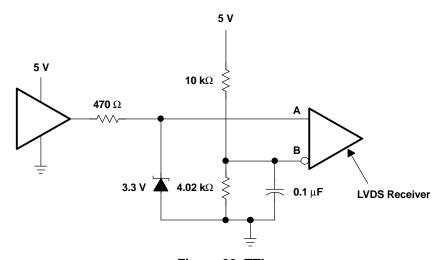


Figure 22. TTL





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#### PACKAGING INFORMATION

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	e Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
SN65LVDS108DBT	ACTIVE	TSSOP	DBT	38	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
SN65LVDS108DBTG4	ACTIVE	TSSOP	DBT	38	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
SN65LVDS108DBTR	ACTIVE	TSSOP	DBT	38	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
SN65LVDS108DBTRG4	ACTIVE	TSSOP	DBT	38	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

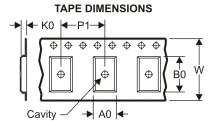
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## TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

## QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	_	Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
SN65LVDS108DBTR	TSSOP	DBT	38	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1





## \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
SN65LVDS108DBTR	TSSOP	DBT	38	2000	346.0	346.0	33.0

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